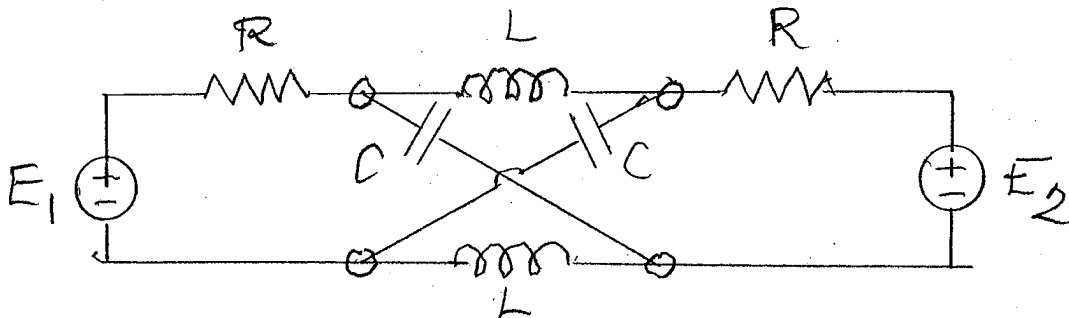


OPEN BOOK - ALL PROBLEMS EQUALLY WEIGHTED

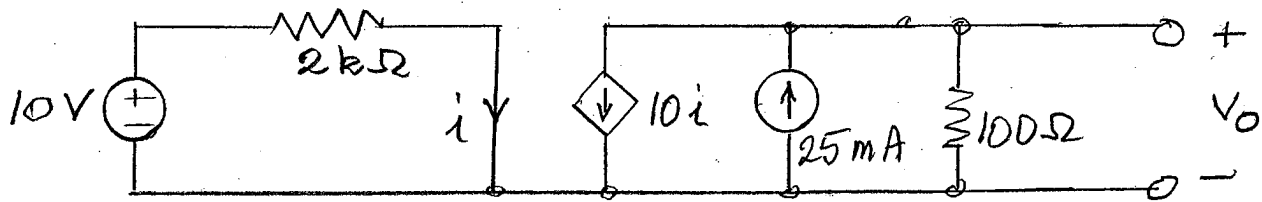
Prof. G. Temes

1. For the terminated two-port shown, $L = 1 \mu\text{H}$ and $C = 4 \text{ pF}$. The scattering parameter S_{11} is equal to 0 for all frequencies.

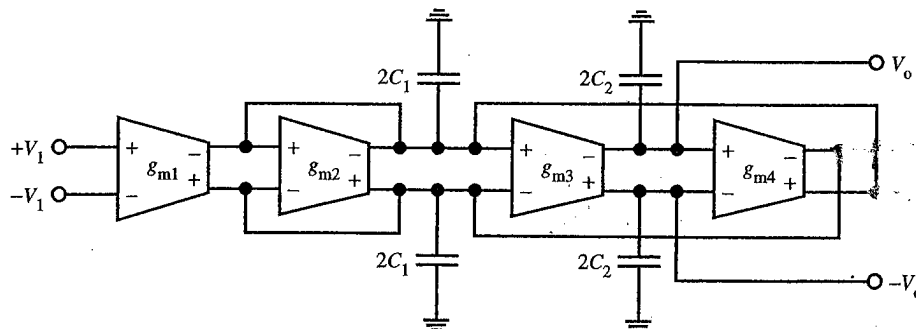
- What are the impedance parameters z_{ij} of the twoport? What is the value of R ?
- What are the absolute values of the other three scattering parameters? Why?
- Derive $S_{12}(s)$ and $S_{21}(s)$!



2. Find the Thevenin equivalent of the circuit shown using the adjoint network method.



3. Find the transfer function of the filter shown.



1. a.

$$V_1 = z_{11} I_1 + z_{12} I_2$$

$$V_2 = -R I_2 = z_{21} I_1 + z_{22} I_2 \rightarrow I_2 = \frac{-z_{21} I_1}{z_{22} + R}$$

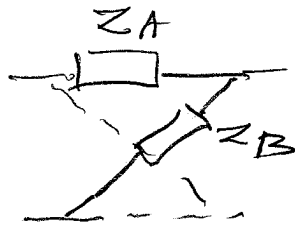
$$V_1 = I_1 \left[z_{11} - \frac{z_{12} z_{21}}{z_{22} + R} \right] = z_{in} I_1$$

$$z_{in} = \frac{z_{11}^2 - z_{12}^2 + R z_{11}}{z_{11} + R} \stackrel{!}{=} R$$

$$z_{11}^2 - z_{12}^2 = R^2 = (z_{11} + z_{12})(z_{11} - z_{12})$$

$$z_{11} = \frac{z_A + z_B}{2}, \quad z_{21} = z_{11} \frac{z_B - z_A}{z_B + z_A} = \frac{z_B - z_A}{2}$$

$$R^2 = z_A z_B = \frac{SL}{SC} = \frac{L}{C} \rightarrow R = \sqrt{\frac{L}{C}} = 500 \Omega$$



$$z_{11} = \frac{s^2 LC + 1}{2sC} = z_{22}$$

$$z_{12} = \frac{1 - s^2 LC}{2sC} = z_{21}$$

$$z_{11}, z_{12} = \frac{1 \pm 4 \cdot 10^{-18} s^2}{8 \times 10^{-12} s}$$

b. By symmetry, $S_{22} \equiv 0$.

Since both ports are matched, and two-port is lossless, $|S_{12}| = |S_{21}| \equiv 1$.

1. c

$$z_{11} = z_{22} = \frac{z_A + z_B}{2}$$

$$z_{12} = z_{21} = \frac{z_B - z_A}{z_B + z_A} z_{11} = \frac{z_B - z_A}{2}$$

$$V_1 = \frac{E_1}{2} = z_{11} \frac{E_1}{2R} - z_{12} \frac{V_2}{R}$$

$$V_2 \frac{z_{12}}{R} = \frac{E_1}{2R} [z_{11} - R]$$

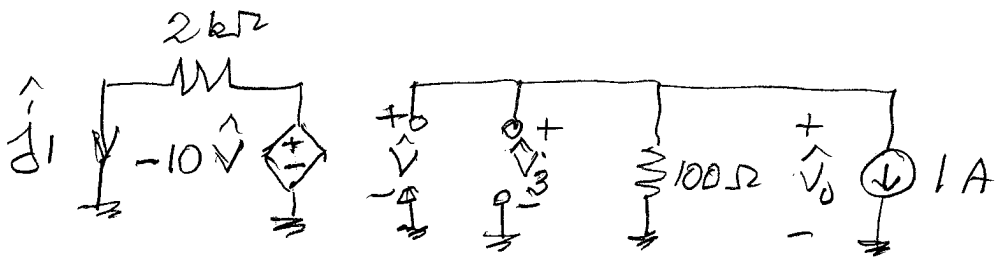
$$S_{21} = \frac{V_2}{E_1/2} = \frac{z_{11} - R}{z_{12}} = \frac{z_A + z_B - 2R}{z_B - z_A} = S_{12}$$

$$= \frac{sL + 1/sC - \sqrt{L/C}}{1/sC - sL} = \frac{s^2 LC - 2s\sqrt{LC} + 1}{1 - s^2 LC}$$

$$S_{12} = \frac{(s\sqrt{LC} - 1)^2}{(1 + s\sqrt{T})(1 - s\sqrt{T})} = \frac{1 - sT}{1 + sT} \quad \begin{array}{l} T \triangleq \sqrt{LC} \\ T = 2ns \end{array}$$

$$|S_{21}(j\omega)| = \left| \frac{1 - j\omega T}{1 + j\omega T} \right| = 1 \quad \forall \omega$$

2. The adjoint network is

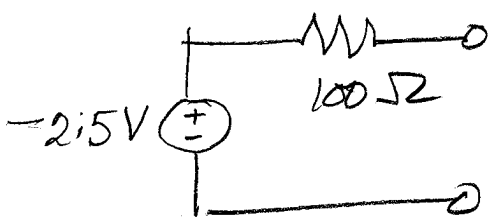


$$\hat{V} = \hat{V}_3 = \hat{V}_0 = -100 \text{ V} \quad , \quad \hat{j}_1 = \frac{+10 \times 100}{2000} = +0.5 \text{ A}$$

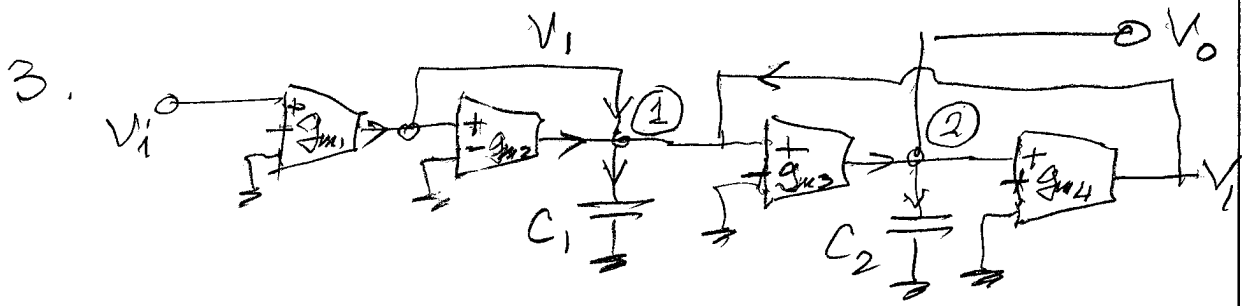
$$V_0 = -\hat{j}_1 e + \hat{V}_3 i = -0.5 \times 10 + (-100)(-0.025) = -2.5 \text{ V}$$

(From eq. 9-87)

$$Z_{Th} = \hat{Z}_{Th} = -\hat{V}_0 = 100 \Omega$$



Easy to check from W!



KCL @ ① :

$$-g_{m1} V_i - g_{m2} V_1 + g_{m4} V_0 = s C_1 V_1$$

$$V_1 = (g_{m4} V_0 - g_{m1} V_i) / (s C_1 + g_{m2})$$

KCL @ ②

$$-g_{m3} V_1 = s C_2 V_0$$

$$g_{m4} V_0 - g_{m1} V_i = (s C_1 + g_{m2}) \frac{s C_2 V_0}{-g_{m3}}$$

$$[g_{m3} g_{m4} + (s C_1 + g_{m2}) s C_2] V_0 = g_{m1} g_{m3} V_i$$

$$\frac{V_0}{V_i} = g_{m1} g_{m3} / (s^2 C_1 C_2 + s g_{m2} C_2 + g_{m3} g_{m4})$$